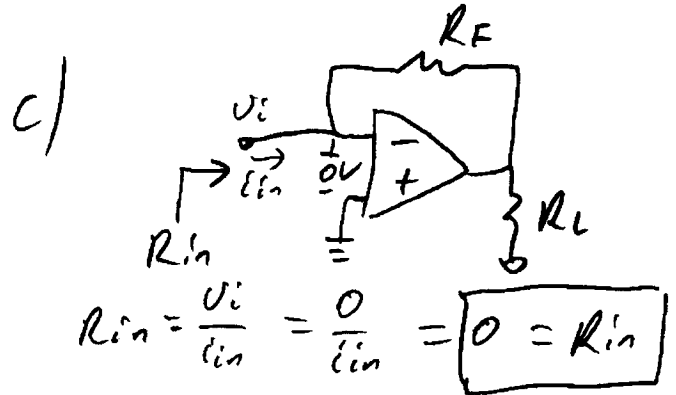
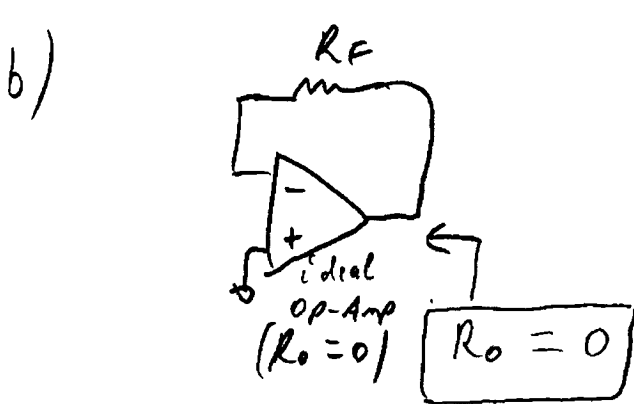


# EECE 2412 : HW2 Solutions Fall 2018

Prob. 1 a)  $i_{in} = \frac{0V - V_o}{R_F} \rightarrow V_o = -R_F \cdot i_{in}$

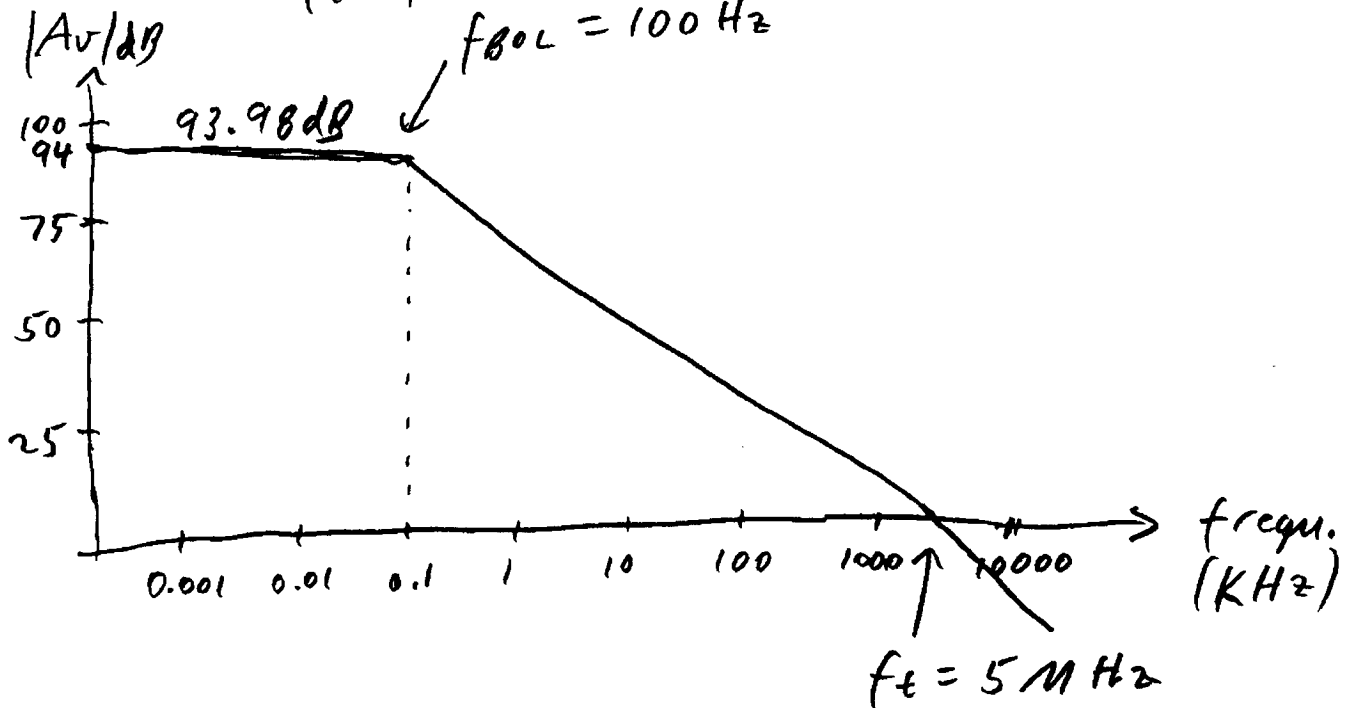


d) ideal transresistance amplifier.

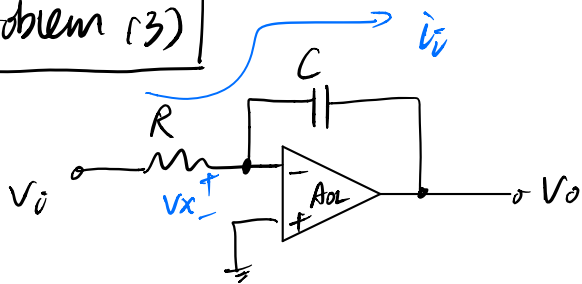
Prob. 2

$20 \cdot \log(50000) = 93.98 \text{ dB}$

$f_t = f_{BOL} \cdot A_{OOL} = 100 \cdot 50000 = 5 \text{ MHz}$



# Problem (3)



$$(a) \quad i_i = \frac{V_i - V_x}{R} = \frac{V_x - V_o}{\frac{1}{sC}} \quad (1)$$

$$V_o = -V_x \cdot A_{OL} \longrightarrow V_x = -\frac{1}{A_{OL}} \cdot V_o$$

apply (2) to (1):

$$\frac{1}{R} (V_i + \frac{1}{A_{OL}} V_o) = sC (-\frac{1}{A_{OL}} V_o - V_o)$$

$$V_o \left[ \frac{1}{R} \cdot \frac{1}{A_{OL}} + sC \left( 1 + \frac{1}{A_{OL}} \right) \right] = -\frac{1}{R} V_i$$

$$A_{CL}(s) = \frac{V_o}{V_i} = \frac{-\frac{1}{R}}{sC \left( 1 + \frac{1}{A_{OL}} \right) + \frac{1}{R} \cdot \frac{1}{A_{OL}}}$$

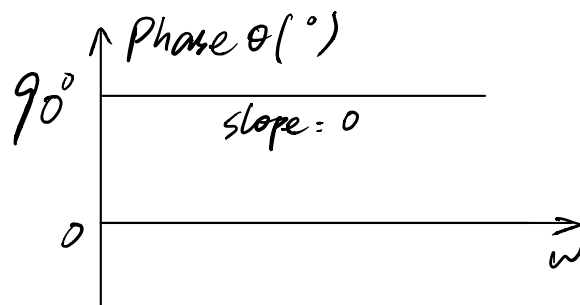
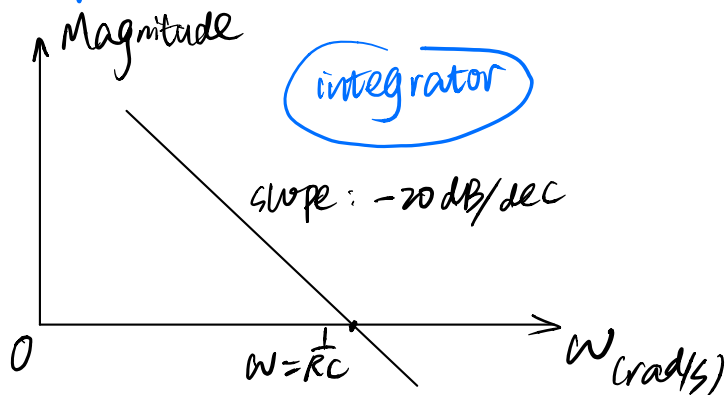
$$= -\frac{\frac{1}{sRC}}{1 + \frac{1}{A_{OL}} + \frac{1}{sRC} \cdot \frac{1}{A_{OL}}}$$

$$A_{CL}(j\omega) = \frac{-\frac{1}{j\omega RC}}{1 + \frac{1}{A_{OL}} + \frac{1}{j\omega RC} \cdot \frac{1}{A_{OL}}}$$

(b)  $A_{OL} \rightarrow \infty$

$$A_{CL}(j\omega) = -\frac{1}{j\omega RC} \quad \begin{cases} 20 \log |A_{CL}(j\omega)| = 20 \log \frac{1}{\omega RC} = -20 \log \omega RC \\ \theta = -\arctan \left( \frac{\omega RC}{0} \right) = -90^\circ \end{cases}$$

NOTE: eq 2.55 in textbook



(c)

$$A_{CL}(j\omega) = \frac{-\frac{1}{j\omega RC}}{1 + \frac{1}{A_{OL}} + \frac{1}{j\omega RC} \cdot \frac{1}{A_{OL}}}$$

$$\omega = 2\pi f = 2\pi \cdot 100 = 200\pi \text{ rad/s}$$

$$A_{OL} = 10^4$$

$$R = 50 \text{ k}\Omega = 50 \times 10^3 \Omega$$

$$C = 50 \text{ pF} = 50 \times 10^{-12} \text{ F}$$

$$A_{CL}(j200\pi) = \frac{-\frac{1}{j200\pi \times 50 \times 10^3 \times 50 \times 10^{-12}}}{1 + \frac{1}{10^4} + \frac{1}{j200\pi \times 50 \times 10^3 \times 50 \times 10^{-12} \times 10^4}}$$

$$= 635.2703 \angle 93.6^\circ$$

[ For comparison, with ideal opamp ( $A_{OL} = \infty$ ) ]

$$A_{CL}(j200\pi) = 636.6198 \angle 90^\circ$$

Output at 100 Hz:

$$V_{out}(j200\pi) = A_{CL}(j200\pi) \cdot V_{in}(j200\pi)$$

$$= (635.2703 \angle 93.6^\circ) \cdot (5 \text{ mV} \angle 0^\circ)$$

$$= 3.176 \angle 93.6^\circ \text{ [V]}$$

$$V_{out}(t) = 3.176 \cdot \sin(2\pi \cdot 100t + 93.6^\circ)$$

$$\text{output amplitude} = 3.176 \text{ V}$$

Problem 4

eqn. 2.39 in the book:  $f_t = A_{ocL} \cdot f_{ocL} = A_{ocL} \cdot f_{ocL}$

$$f_{3dB} = f_{ocL} = \frac{f_t}{A_{ocL}} = \frac{20 \times 10^6 \text{ Hz}}{100} = \boxed{200 \text{ kHz} = f_{ocL}}$$

Eqn. 2.37 in the book provides the closed-loop response:

$$A_{cL}(f) = \frac{A_{ocL}}{1 + j \frac{f}{f_{ocL}}} = \frac{100}{1 + j \frac{f}{200 \times 10^3}} \quad \text{Phase:} \rightarrow \angle A_{cL}(f) = \tan^{-1}\left(\frac{0}{100}\right) - \tan^{-1}\left(\frac{f/f_{ocL}}{1}\right)$$

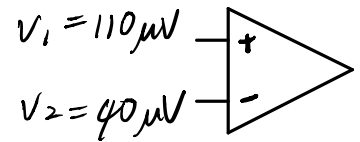
$$\angle A_{cL}(f) = 0 - \tan^{-1}\left(\frac{f/f_{ocL}}{1}\right) = -\tan^{-1}\left(\frac{f}{f_{ocL}}\right)$$

$$-6^\circ = -\tan^{-1}\left(\frac{f_1}{200 \times 10^3}\right) \rightarrow \boxed{f_1 = 21.02 \text{ kHz}} \quad (-6^\circ \text{ phase shift})$$

$$-84^\circ = -\tan^{-1}\left(\frac{f_2}{200 \times 10^3}\right) \rightarrow \boxed{f_2 = 1.903 \text{ MHz}} \quad (-84^\circ \text{ phase shift})$$

# Problem (5)

$$CMRR = \frac{|A_{dm}|}{|A_{cm}|} = 85 \text{ dB}, \quad |A_{dm}| = 102 \text{ dB}$$



$$(a) \quad V_{in, dm} = V_1 - V_2 = 110 \mu\text{V} - 40 \mu\text{V} = 70 \mu\text{V}$$

$$(b) \quad V_{in, cm} = \frac{1}{2}(V_1 + V_2) = \frac{1}{2}(110 \mu\text{V} + 40 \mu\text{V}) = 75 \mu\text{V}$$

$$(c) \quad |A_{dm}| = 102 \text{ dB} \rightarrow 10^{\left(\frac{102}{20}\right)} = 125892.54$$

$$CMRR = 85 \text{ dB} \rightarrow 10^{\left(\frac{85}{20}\right)} = 17782.79$$

$$|A_{cm}| = \frac{CMRR}{|A_{dm}|} = 102 - 85 = 17 \text{ dB} \rightarrow 10^{\left(\frac{17}{20}\right)} = 7.09$$

$$|A_{cm}| = 7.09$$

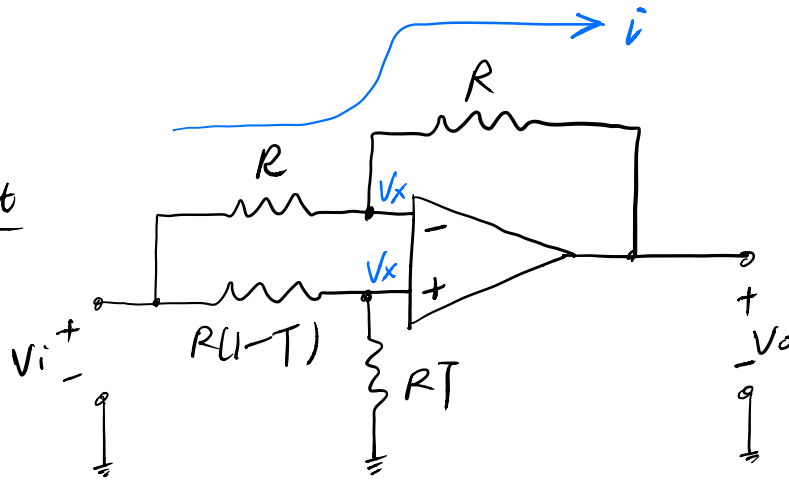
$$V_{out} = V_{in, cm} |A_{cm}| + V_{in, dm} |A_{dm}|$$

$$= 75 \mu\text{V} \times 7.09 + 70 \mu\text{V} \times 125892.54$$

$$V_{out} = 8.81301 \text{ V}$$

# Problem (b)

equivalent circuit



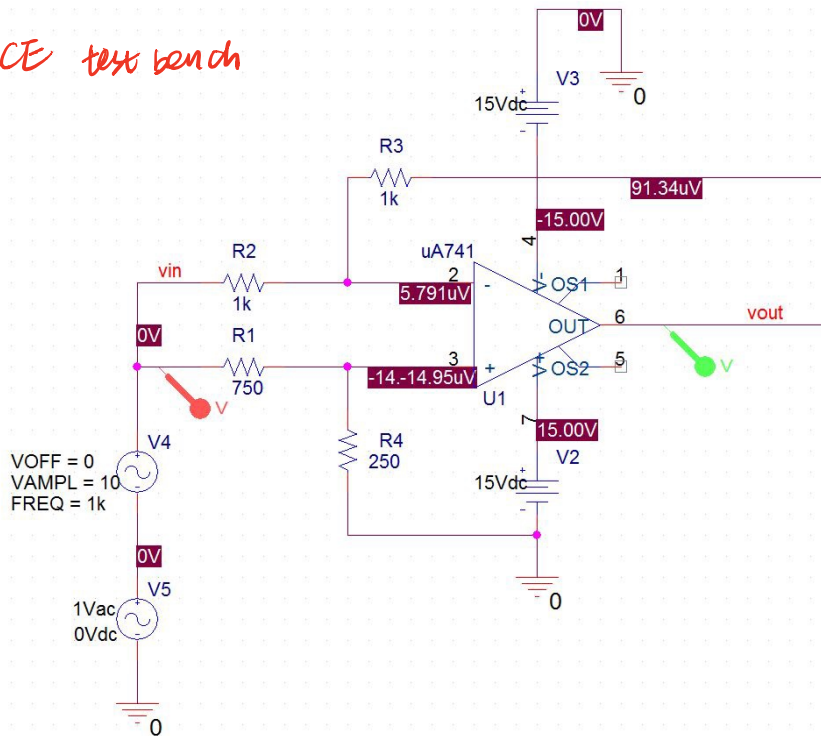
$$\begin{cases} i = \frac{V_i - V_x}{R} = \frac{V_x - V_o}{R} \\ V_x = \frac{R_T}{R} \cdot V_i = T \cdot V_i \end{cases}$$

$$V_i - T V_i = T V_i - V_o$$

$$V_o = (2T - 1) \cdot V_i$$

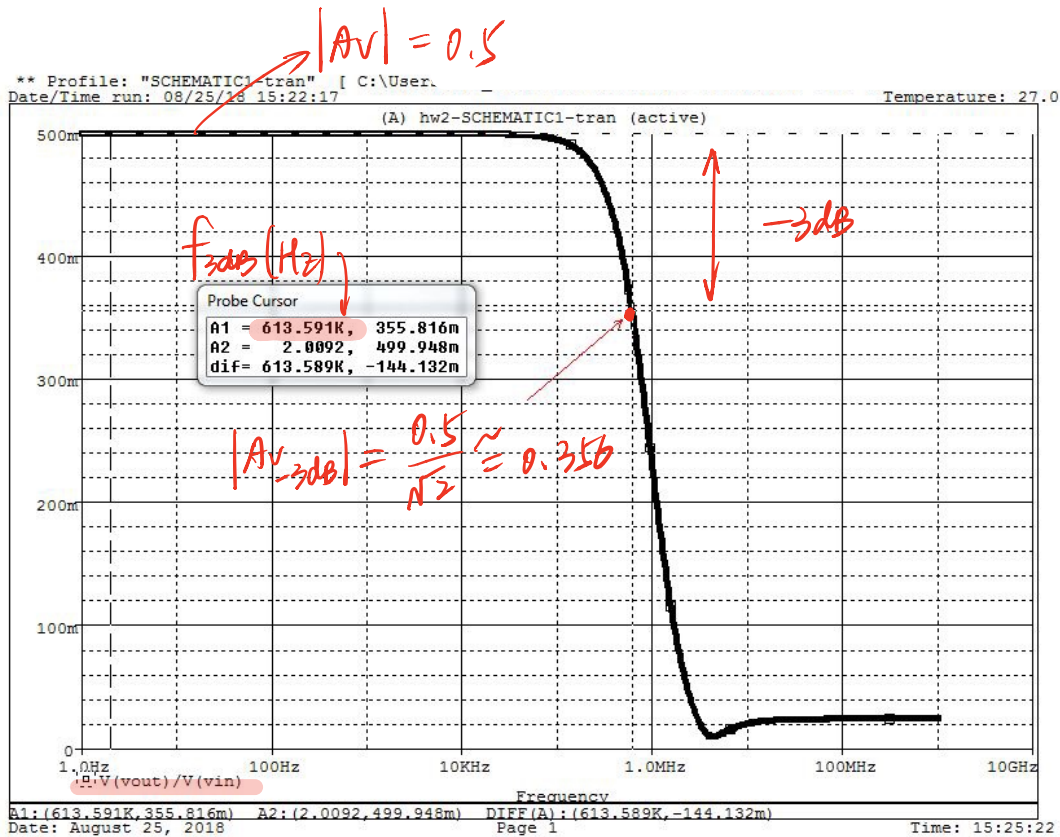
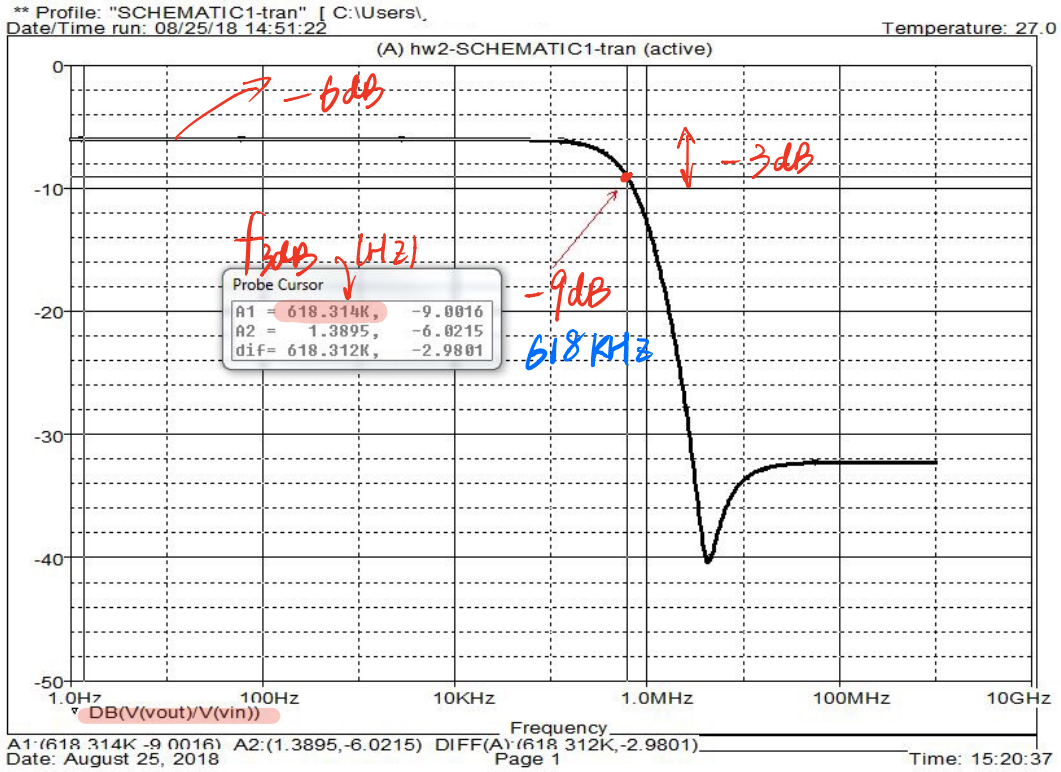
$$A_V = \frac{V_o}{V_i} = 2T - 1 \quad \xrightarrow{T=0.25} \quad A_V = -0.5 \rightarrow \boxed{-6\text{dB}}$$

PSPICE test bench

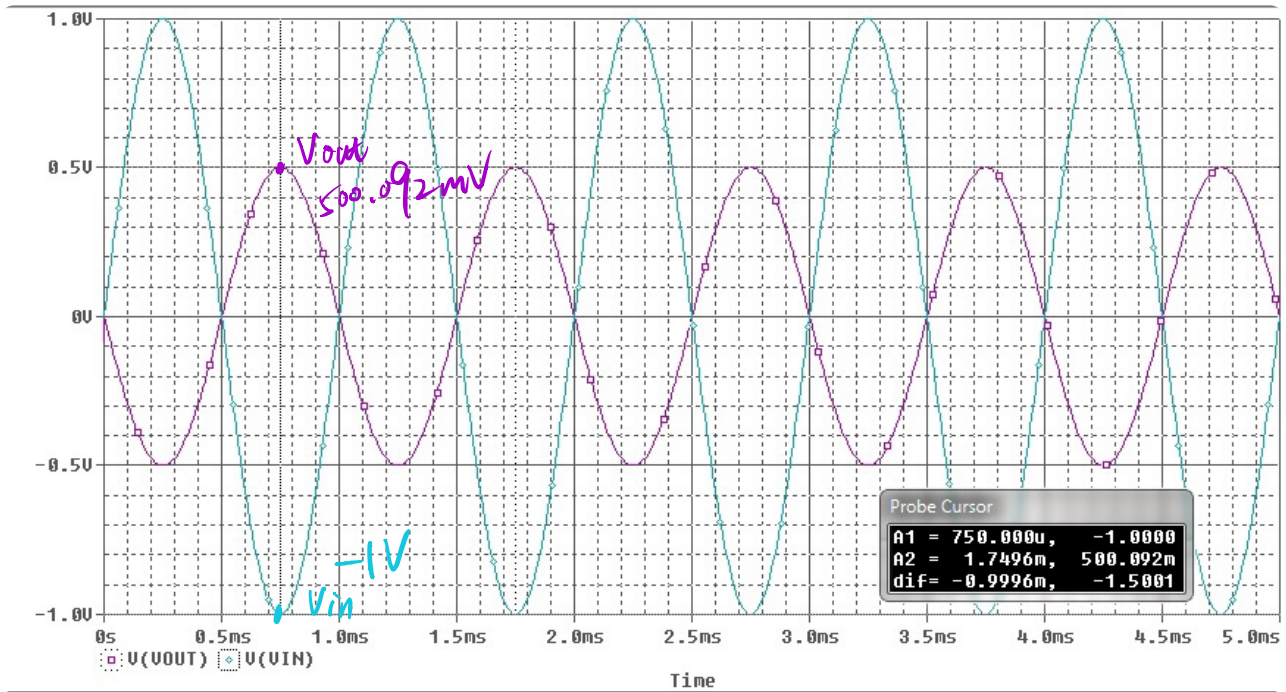


(a)

### Voltage gain in dB vs. frequency



(b) Transient input/output voltages with a sinusoid always input signal with an amplitude of 1V @ 1kHz

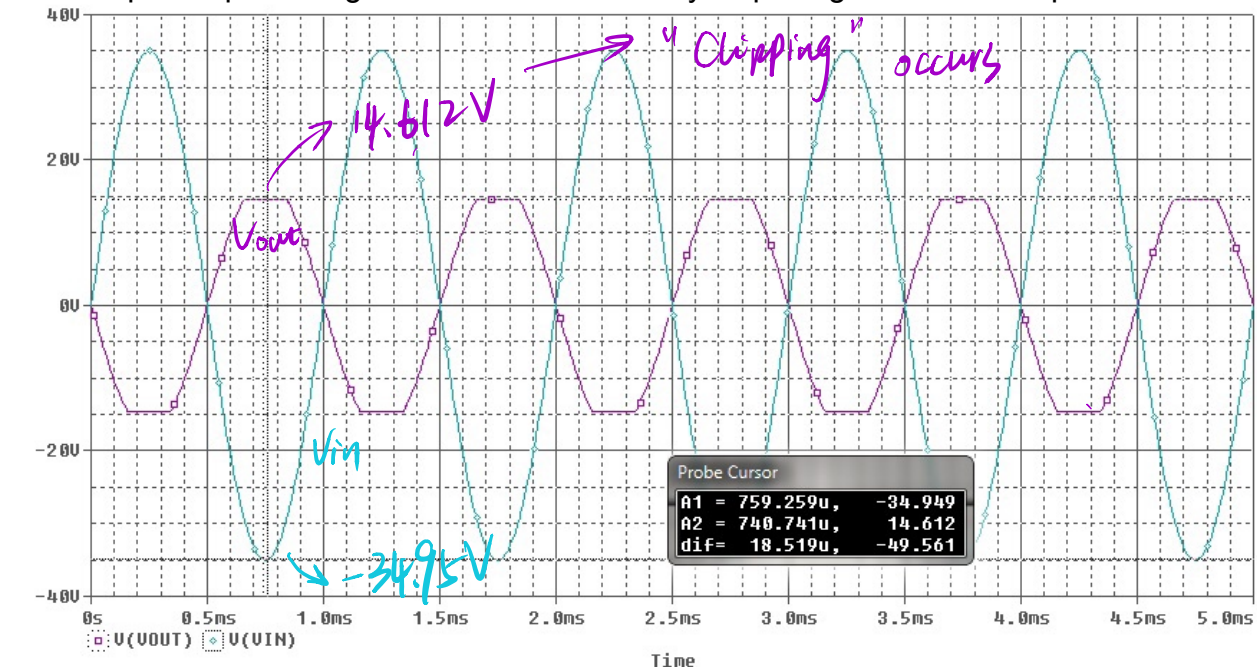


$$A_v = \frac{-1V}{0.500092V} = -0.50092$$

$$\text{expected } A_v = 2T - 1 = -0.5$$

Theory and simulation results match.

(c) Transient input/output voltages with a sinusoid always input signal with an amplitude of 35V @ 1kHz

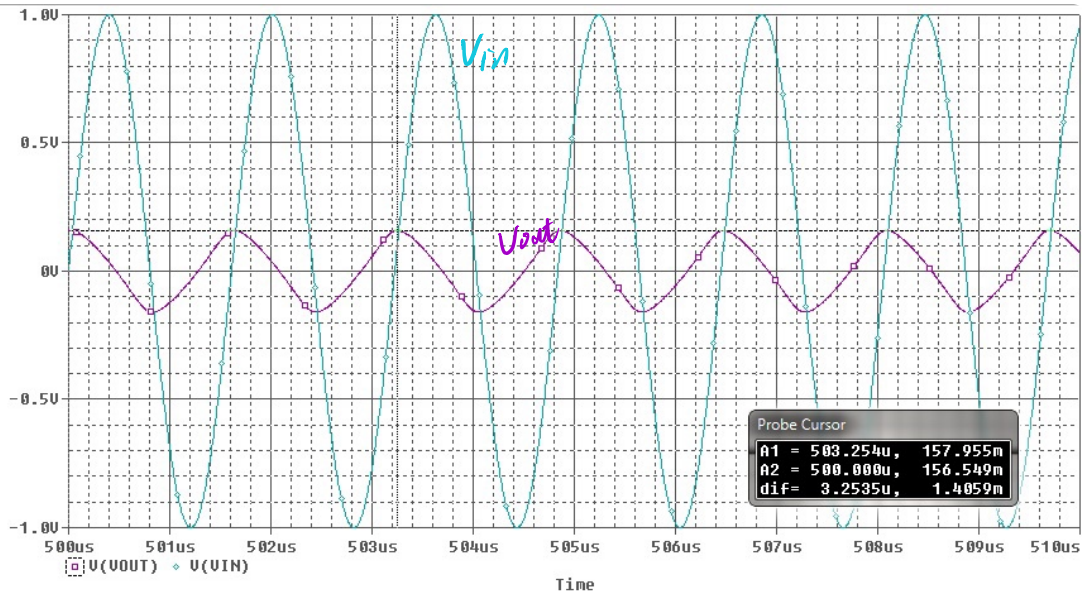


Signal "vout" goes clipped as a result of "signal can never reach beyond the supply voltage ( $\pm 15V$ )"

↓ saturation

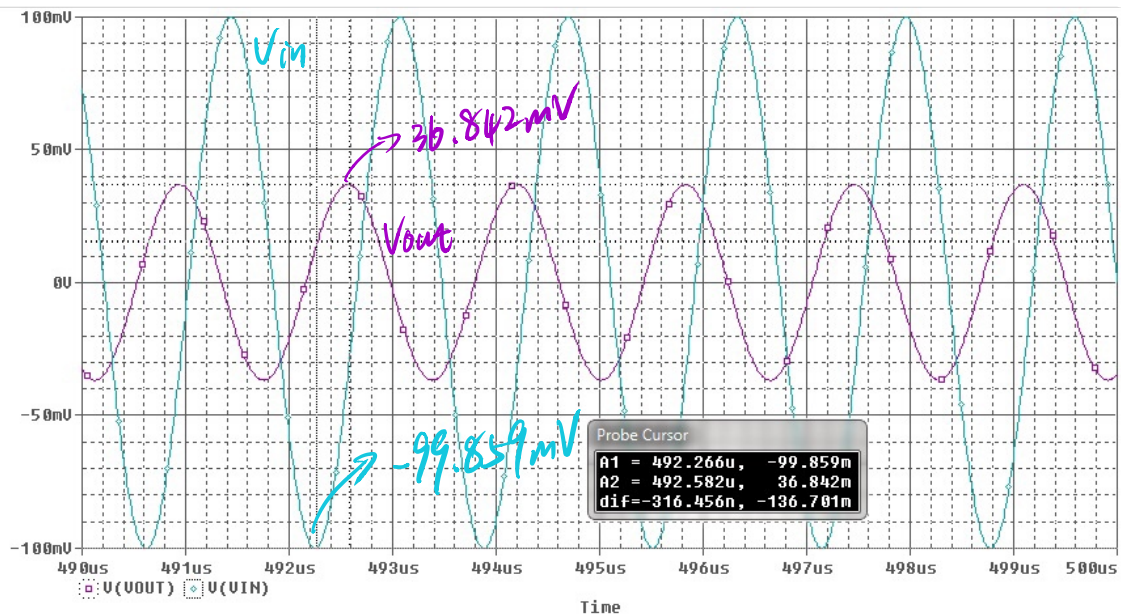


(d) Transient input/output voltages with a sinusoid always input signal with an amplitude of 1V @ 613kHz



\* The output signal is almost equal to triangle wave as a result of the slew rate limiting from the op-amp.

(e) Transient input/output voltages with a sinusoid always input signal with an amplitude of 0.1V @ 613kHz



$$V_{in(p-p)} = 99.718 \text{ mV}$$

$$V_{out(p-p)} = 73.684 \text{ mV}$$

$$A_v = \frac{V_{out(p-p)}}{V_{in(p-p)}} \approx 0.369$$

NOTES

$$f_{3dB} = 613 \text{ kHz}$$

$$A_{3dB} = \frac{A_{DC}}{\sqrt{2}} = \frac{0.5}{\sqrt{2}} = 0.353$$

Simulation results agree with theoretical expectation