# EECE 2413 Electronics Laboratory

#### Lab #3: Bipolar Junction Transistor Basics

# <u>Goals</u>

In this lab you will experiment with the *bipolar junction transistor* or the BJT. Initially you will bias a BJT in the three most common regions: active, saturation, and cut-off. In the active region you will determine the DC current gain of the transistor ( $\beta$ ) and discover that  $\beta$  varies from BJT-to-BJT and with temperature.

You will learn about current gain by building a "touch switch" which lights an LED as you touch two low-current electrodes. A low-level current (a few microamperes) passing through your body is *amplified* by the BJT to a value that is sufficient to drive an LED (a milliampere or so).

As always, take your time during these experiments. Think about *what* you are being asked to do and *why* the experiments are important.

Once you are comfortable with the way BJTs work, you will design and test your own *automatic night light* using a light-sensitive resistor (CdS photocell), a BJT, an LED and assorted resistors.

# <u>Prelab</u>

Prelabs will be collected for grading at the beginning of the lab. Keep a copy for your own use during the lab!

1. The voltage measured between the collector and ground in Figure 1 is 3.72 volts. What is the DC current gain ( $\beta$ ) of the BJT in this circuit? (Assume V<sub>BE</sub> = 0.7 V)

2. What is the smallest value of the  $R_B$  which will allow the BJT in Figure 1 to remain in the forward active region ( $V_C > V_B$ )? As  $R_B$  becomes smaller than this value, what region is the BJT operating in? (Hint: first find  $I_C$  and  $I_B$ )

3. You can force the BJT in Figure 1 into cutoff by adding another resistor from the <u>base</u> to <u>ground</u>. Assuming that the BJT is cutoff if  $V_{BE} \le 0.3$  volts, find the *maximum* value of this new resistor that will cause the BJT to operate in the cutoff region.

4. Read Part 2 (a) and sketch the circuit you think will work. Also draw the modified circuit as described in part 2 (b). You may change your design as you work through the lab, however. (You do not need to draw the Optional Design Circuits 1 and 2)

5. Read the lab experiment and see the instructor with any questions you may have.

#### Part 1: Basic BJT behavior

You have two 2N3904 BJTs in your parts kit. So that you can distinguish between them, cut the leads on one of the BJTs a *little* bit shorter. Name each BJT ( $Q_1$  and  $Q_2$ ) and note in your lab book which BJT has the shorter leads. Construct the circuit shown in Figure 1. Measure and record the values of both resistors, as usual. *Note:* Sometimes the emitter, base, and collector leads are labeled on the package or case of the BJT. If they are not labeled, refer to Figure 2.



Figure 1. Basic BJT Biasing Circuit  $R_B = 309 \text{ k}\Omega, R_C = 1 \text{ k}\Omega$ 



Figure 2. View of 2N3904 npn BJT

#### Concept: Forward active region of the BJT and the variation of $\beta$

The forward active region of the BJT is most commonly used for amplifier circuits. This is because the BJT behaves nearly linearly ( $i_c = \beta i_B$ ) in the forward active region. In Lab 3 you will experiment with BJT amplifiers. For now, let's just look at the BJT biased in the forward active region.

a. Measure and record  $V_{CE}$  and  $V_{BE}$  for both transistors. Explain why the BJTs are biased in the forward active region in terms of  $V_{BE}$  and  $V_{CE}$ .

b. Calculate  $I_B$ ,  $I_C$ , and  $\beta$  for each transistor from the data of part 1a.

c. A 2N3904 is setup on the curve tracer in the lab. From the i-v characteristics displayed on the curve tracer, find  $\beta$  for this device under the same bias conditions as part 1a. Ask the TA or Professor for help using the curver tracer!

d. Ask two other lab groups for their two values of  $\beta$ . Calculate the average and the standard deviation of the seven  $\beta$ 's. Discuss the problem you would have designing a circuit that *required*  $\beta$ =150 ±1%. Is this practical? In the data sheets at the end of the lab, h<sub>FE</sub> is used as the symbol for  $\beta$ . Are all 7 of these transistors within specs?

*Note:* If  $<\beta>$  is the average value of  $\beta_1$ ,  $\beta_2$ , and  $\beta_3$ , then the standard deviation is found

from  $\sigma = \sqrt{\frac{1}{n} \sum_{i=1}^{n} (\beta_i - \langle \beta \rangle)^2}$  where n=3. You will use n=7.

e. Get the can of refrigerant from the instructor and cool the BJT in Fig.1 using a 1 second blast. Have your lab partner measure the maximum value of  $V_{CE}$ . Repeat the experiment and determine max[ $V_{BE}$ ]. Calculate  $\beta$  for this BJT when it is cold. Is this transistor within specification (see attached data sheets)? Would you want this circuit in your car stereo on a cold winter morning? Discuss this in your lab report.

In this experiment so far you have determined that  $\beta$  varies from BJT to BJT.  $\beta$  also varies with temperature. Remember this! It is so important that the **first graph** in the spec sheets show the typical temperature variation. Later in electronics you will learn to design circuits which function more or less independently of  $\beta$  and therefore independent of temperature.

# Concept: BJT saturation - a switch in the "on" position

In *saturation*, the BJT is acting like a switch in the "on" position. That is, the collector and emitter are nearly shorted together so the  $V_{CE}$  is approximately 0.1-0.3 volts.

f. Replace the 309 k $\Omega$  resistor in Figure 1 with another resistor which forces the BJT into the saturation region. Refer to the Prelab, question 2. Write the value of R<sub>B</sub> you have chosen in your lab book and **describe your choice in the lab report**.

g. Measure  $V_{CE(SAT)}$  and  $V_{BE}$  for both 2N3904's. How much variation do you observe? How important is BJT-to-BJT variation if the saturation region is used?

h. Calculate the current gain  $(I_C/I_B)$  for each of the BJTs when saturated. Compare this number with  $\beta$  in the forward active region. If a high  $\beta$  is important, which region should you use?

# Concept: BJT cutoff - a switch in the "off" position

In cutoff the BJT is acting like a switch in the "off" position. That is, the path from collector to emitter is very nearly an open circuit.

- i. There are a few ways to achieve cutoff. Try *each* of these and **record**  $V_{CE}$ .
  - 1) remove the base resistor so that  $I_B = 0$
  - 2) short the base to ground so that  $V_{BE} = 0$
  - 3) put a small resistor between base and ground such that  $V_{BE} \ll 0.7 v$ .

In part (3) above, use the result from your prelab. Note that this technique is important to understand since you may want to use it in the design section (Part 2).

# Concept: A practical application of BJT current gain (a touch sensitive switch)

Using the BJT with the highest  $\beta$ , build the circuit in Figure 3. At the two terminals marked "touch points", insert two bared wires into your protoboard in a place where they can be easily touched with your finger.

The circuit lights the LED as you touch the two low-current electrodes. The low-level current (a few microamperes) passing through your finger is *amplified* by the BJT to a value that is sufficient to drive an LED (a milliampere or so).

*Note:* The resistance of your finger depends on a number of factors. Among them is the level of stress you are under. This is one of the things a "lie detector" might measure! If you are having trouble getting the LED to light, try moistening your finger.

j. By measuring the appropriate node voltages, determine  $I_C$ ,  $I_B$ , and the resistance of your finger (without directly measuring your finger with an ohmmeter!).

k. If you completed a circuit --consisting only of an LED and a 10 V source-with your finger, how much current would flow through the LED (see Fig. 4)? Would the LED emit visible light? Comment on the importance of the BJT in the circuit shown in Fig. 3.



Figure 3. A touch sensitive switch:  $R_C = 1 k\Omega$ 



Figure 4. Non-amplified touch switch: Will it work?

You are now finished with Part 1. Remember that the lab report is due 1 week from today.

#### Part 2: Designing an automatic night light

(a) You will need a light-sensitive resistor for this design. A cadmium sulfide (CdS) photocell has a low resistance when exposed to light (typically 1-4 k) and a high resistance in darkness (typ. 50k-100 k). Using the CdS cell provided, measure and record the light and dark resistances in your lab notebook.

Now design a BJT circuit which turns a red LED *on* when the room lights are *off*. Conversely, the LED should extinguish when the room is lit again. (Cup your hands around the circuit to simulate darkness.)

(b) Once your circuit is working, you will notice that it is susceptible to flicker when the light is momentarily interrupted (for example, if auto headlights were to occasionally illuminate the sensor of an automated street light). This is definitely undesirable. **Modify** your circuit so that the illumination must remain either light or dark for about 1 second before a change in the LED status occurs. *Hint:* You will need to use a capacitor.

<u>Shop Note</u>: Electrolytic capacitors are polarized...that is, they have + and - terminals. If you connect the capacitor with reversed polarity, it will "leak" a small but devastating DC current through its terminals. Always discharge the capacitor *before* inserting it into the circuit. This prevents stored charge in the capacitor from destroying your work.

When your design is working, draw a schematic of the circuit, demonstrate it to the instructor, and get the instructor's *design approval* on your schematic.

# In your lab notebook, describe the operation of your design, the rationale of picking the devices, and include a schematic with the instructor's design approval. All designs will be graded when you hand-in your notebook at the end of the semester.

<u>Optional Design1:</u> Modify your circuit so that the CdS photocell must be illuminated for 1 second before the light emitting diode goes off AND the photocell must be in the dark for 1 second before the LED turns on.

<u>Optional Design2</u>: Add a BJT inverter to your circuit so that an additional green LED is lit when the room lights are "on" but extinguished when the room is darkened. Make certain the red LED still functions as specified in the original design.

JH10.96 rev.1.97 rev.8.05 rev.7.08 rev.6.10

#### Equipment List -- Lab #3

Note: "\*" indicates this component was used in Lab 2.

Agilent E3647A dual output power supply \* Fluke 8010A digital multimeter \* Fluke 45 Dual Display Multimeter \* Proto-Board model PB-103 \*

4 banana plug-terminated test leads \* #20 hook up wire \* wire strippers \* Aerosol component coolant (non-CFC) \*

Diodes:

Red LED (1)\*

Transistors:

2N3904 npn (2)

CdS photocell  $\sim 1 \text{ k}\Omega - 100 \text{ k}\Omega$  range

Resistors: 1/4 W unless otherwise specified

| ±5% | (1) *   |
|-----|---|
| ±5% | (1) *   |
| ±5% | (1) *   |
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|     | $\pm 5\%$<br>$\pm 5\%$<br>$\pm 5\%$<br>$\pm 5\%$<br>$\pm 5\%$<br>$\pm 5\%$<br>$\pm 5\%$ |

Electrolytic capacitors:  $10 \ \mu\text{F}$  to  $1000 \ \mu\text{F}$ ,  $10 \ WVDC$  or greater

#### From the instructor:

A Curve Tracer, set up with a 2N3904 BJT.



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|--|---|------------------------|---|----------|-------------|-----|
|  | DFF CHARACTERISTI<br>V <sub>(BR CE0</sub> Collector-Emi<br>Voltage<br>V <sub>(BR CB0</sub> Collector-Base<br>V <sub>(BR CB0</sub> Emitter-Base<br>I <sub>BL</sub> Base Cutoff C | CS<br>itter Breakdown  | lc = 1.0 mA. la = 0   |          |             |     |
| Vigence Collector-Emitter Breakdown Ic = 1.0 mA, Ig = 0 40 V   Vigence Collector-Base Breakdown Voltage Ic = 10 $\mu$ A, Ig = 0 60 V   Vigence Emitter-Base Breakdown Voltage Ig = 10 $\mu$ A, Ig = 0 60 V   Vigence Emitter-Base Breakdown Voltage Ig = 10 $\mu$ A, Ig = 0 6.0 V   Vigence Emitter-Base Breakdown Voltage Ig = 10 $\mu$ A, Ig = 0 6.0 V   Vigence Emitter-Base Breakdown Voltage Ig = 10 $\mu$ A, Ig = 0 6.0 V   Vigence Collector Cutoff Current Vcg = 30 V, Vgg = 3V 50 nA   Collector Cutoff Current Vcg = 0.1 mA, Vcg = 1.0 V 100 300 10   Vigence DC Current Gain Ic = 0.1 mA, Vcg = 1.0 V 100 300 10   Vigence Collector-Emitter Saturation Voltage Ic = 10 mA, Ig = 1.0 mA 0.2 V 1c = 10 mA, Ig = 1.0 mA 0.2 V   Value Collector-Emitter Saturation Voltage Ic = 10 mA, Ig = 1.0 mA 0.65 0.85 V   Vigence Coll mA, Ig  | V <sub>(BR)CEO</sub> Collector-Emi<br>Voltage<br>V <sub>(BR)CBO</sub> Collector-Bas<br>V <sub>(BR)EBO</sub> Emitter-Base<br>I <sub>BL</sub> Base Cutoff C                       | itter Breakdown        | $l_{c} = 1.0 \text{ mA}$ . $l_{B} = 0$  |          |             |     |
| Vignices Collector-Base Breakdown Voltage I <sub>c</sub> = 10 $\mu$ A, I <sub>E</sub> = 0 60 V   Vignices Emitter-Base Breakdown Voltage I <sub>E</sub> = 10 $\mu$ A, I <sub>c</sub> = 0 6.0 V   aL Base Cutoff Current Voe = 30 V, Vee = 3V 50 nA   cex Collector Cutoff Current Voe = 30 V, Vee = 3V 50 nA   cex Collector Cutoff Current Voe = 30 V, Vee = 3V 50 nA   DN CHARACTERISTICS* Ic = 0.1 mA, Voe = 1.0 V 40 70 100 300   lo = 50 mA, Voe = 1.0 V 100 100 300 Ic = 10 mA, Voe = 1.0 V 60 100 100 300 Ic = 10 mA, Voe = 1.0 V 70 100 300 Ic = 10 mA, Voe = 1.0 V 100 300 Ic = 10 mA, Voe = 1.0 V 30 Ic = 10 mA, Voe = 1.0 V 30 Ic = 10 mA, Voe = 1.0 V 30 Ic = 10 mA, Voe = 1.0 V 30 Ic = 10 mA, Ie = 1.0 mA 0.5 Voe = 3.0 V, Vee = 0.5 V, Ie = 0, Ic = 1.0 mA, Ie = 5.0 mA Ic = 50 mA, Ie = 5.0 mA Ic = 10 mA, Ie = 1.0 mA Ic = 1.0 MHz Ic = 10 MA, Ic = 1.0 MA Ic = 1.0 MA Ic = 1.0 MA Ic = 1.0 M   | V(BR)CBO Collector-Bas<br>V(BR)EBO Emitter-Base<br>BL Base Cutoff C   |                        | is the more that is   | 40       |             | V   |
|  | /(BR)EBO Emitter-Base<br>BL Base Cutoff (   | e Breakdown ∀oltage    | I <sub>c</sub> = 10 μA, I <sub>E</sub> = 0  | 60       |             | V   |
| Base Cutoff Current Vor = 30 V, Ver = 3V 50 nA   Correct Cutoff Current Vor = 30 V, Ver = 3V 50 nA   DN CHARACTERISTICS* Ic = 0.1 mA, Vor = 1.0 V 40 na   Pre DC Current Gain Ic = 0.1 mA, Vor = 1.0 V 40 na   Ic = 10 mA, Vor = 1.0 V 100 300 60 100 300 100 300 100 300 100 300 100 300 100 300 100 300 100 300 100 300 100 300 100 300 100 300 100 300 100 300 100 300 100 100 300 100 100 300 100 100 100 300 100  | BL Base Cutoff (  | Breakdown Voltage      | I <sub>E</sub> = 10 μA, I <sub>C</sub> = 0  | 6.0      | <u> </u>    | V   |
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|  | Collector Cut   | off Current            | $V_{CE}$ = 30 V, $V_{EB}$ = 3V  |          | 50          | nA  |
| $ \begin{array}{c c c c c c c c c c c c c c c c c c c $  |   |                        | •   |          | •           | •   |
|  | N CHARACTERISTIC  | S*                     |   |          |             | ,   |
| $\begin{array}{ c c c c c c c c c c c c c c c c c c c$   | FE DC Current G   | ain                    | l <sub>c</sub> = 0.1 mA, V <sub>ce</sub> = 1.0 V<br>l <sub>z</sub> = 1.0 mA V <sub>ce</sub> = 1.0 V | 40<br>70 |             |     |
| $\begin{tabular}{ c  c  c  c  c  c  c  c  c  c  c  c  c $  |   |                        | $I_{c} = 10 \text{ mA}, V_{ce} = 1.0 \text{ V}$   | 100      | 300         |     |
| Ic 10 mA, Vor 10 mA, Is 30 Ic   Cellsati Collector-Emitter Saturation Voltage Ic 10 mA, Is 10 mA,  |   |                        | I <sub>c</sub> = 50 mA, V <sub>ce</sub> = 1.0 V   | 60       |             |     |
| Cellise() Collector-Emitter Saturation Voltage I <sub>0</sub> = 10 mA, I <sub>8</sub> = 1.0 mA 0.2 V   I <sub>0</sub> = 50 mA, I <sub>8</sub> = 5.0 mA 0.3 V   I <sub>0</sub> = 50 mA, I <sub>8</sub> = 5.0 mA 0.65 0.85 V   I <sub>0</sub> = 50 mA, I <sub>8</sub> = 5.0 mA 0.65 0.85 V   MALL SIGNAL CHARACTERISTICS I <sub>0</sub> = 10 mA, I <sub>8</sub> = 5.0 mA 0.95 V   MALL SIGNAL CHARACTERISTICS I <sub>0</sub> = 10 mA, V <sub>0E</sub> = 20 V, f = 100 MHz 300 MHz   Voite V <sub>0</sub> = 5.0 V, I <sub>8</sub> = 0, f = 1.0 MHz 4.0 pF   Voite I = 1.0 MHz 4.0 pF   Vibio Input Capacitance V <sub>E8</sub> = 0.5 V, I <sub>0</sub> = 0, f = 1.0 MHz 8.0 pF   Vibio Input Capacitance V <sub>E8</sub> = 0.5 V, I <sub>0</sub> = 0, f = 1.0 MHz 5.0 dB   WITCHING CHARACTERISTICS I <sub>0</sub> = 100 µA, V <sub>0E</sub> = 5.0 V, I <sub>8</sub> = 0.5 V, I <sub>8</sub> = 0.5 V, I <sub>8</sub> = 0.5 V, I <sub>9</sub> = 1.0 kQ, f=10 Hz to 15.7 kHz 35 ns   WITCHING CHARACTERISTICS V <sub>0C</sub> = 3.0 V, V <sub>8E</sub> = 0.5 V, I <sub>9</sub> = 10 mA 35 ns   Storage Time I <sub>0</sub> = 10 mA, I <sub>81</sub> = 1.0 mA 50 ns   Fall Time I <sub>2</sub> = 1.0 mA  | Oslisster Freit   | Columbian Maltana      | I <sub>c</sub> = 100 mA, V <sub>CE</sub> = 1.0 V  | 30       | 0.0         |     |
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| Ic = 50 mA, Ig = 5.0 mA 0.95 V   MALL SIGNAL CHARACTERISTICS r Current Gain - Bandwidth Product Ic = 10 mA, Vcg = 20 V, f = 100 MHz 300 MHz   Note Output Capacitance Vcg = 5.0 V, Ig = 0, f = 1.0 MHz 4.0 pF   Note Input Capacitance Vcg = 5.0 V, Ig = 0, f = 1.0 MHz 8.0 pF   Note Input Capacitance Vcg = 0.5 V, Ic = 0, f = 1.0 MHz 8.0 pF   IF Noise Figure Ic = 100 µA, Vcg = 5.0 V, g = 0.5 V, Ic = 0, f = 1.0 MHz 5.0 dB   WITCHING CHARACTERISTICS Vcc = 3.0 V, Vg = 0.5 V, Ic = 0.5 V, Ic = 0.5 V, Ic = 0.5 V, Ic = 10 mA, Ig = 1.0 mA 35 ns   Storage Time Vcc = 3.0 V, Vg = 0.5 V, Ic = 10 mA 35 ns   Ic = 10 mA, Ig = 1.0 mA 35 ns 1c = 10 mA, Ig = 1.0 mA 35 ns   | Base-Emitter  | Saturation Voltage     | $I_c = 10 \text{ mA}, I_B = 1.0 \text{ mA}$   | 0.65     | 0.85        | Ň   |
| MALL SIGNAL CHARACTERISTICS   r Current Gain - Bandwidth Product $I_0 = 10 \text{ mA}, V_{CE} = 20 \text{ V}, \\ f = 100 \text{ MHz} \\ V_{CE} = 5.0 \text{ V}, I_E = 0, \\ f = 1.0 \text{ MHz} \\ V_{CE} = 5.0 \text{ V}, I_E = 0, \\ f = 1.0 \text{ MHz} \\ V_{EB} = 0.5 \text{ V}, I_C = 0, \\ f = 1.0 \text{ MHz} \\ IF$ Noise Figure $V_{EB} = 0.5 \text{ V}, I_C = 0, \\ f = 1.0 \text{ MHz} \\ I_C = 100 \ \mu\text{A}, V_{OE} = 5.0 \text{ V}, \\ R_0 = 1.0 \text{ KQ}, f = 10 \text{ Hz} \text{ to } 15.7 \text{ KHz} \\ V_{CC} = 3.0 \text{ V}, V_{BE} = 0.5 \text{ V}, \\ IC = 100 \ \mu\text{A}, I_{B1} = 1.0 \text{ mA} \\ I_C = 100 \ \mu\text{A}, I_{B1} = 1.0 \text{ mA} \\ I_C = 100 \ \mu\text{A}, I_{B1} = 1.0 \text{ mA} \\ I_C = 100 \ \mu\text{A}, I_C = 100 \ \mu\text{A} \\ I_C = 100 \ \mu\text{A}, I_C = 100 \ \mu\text{A} \\ I_C = 1000 \ \mu\text{A} \\ I_C = 10000 \ \mu\text{A} \\ I_C$ |   |                        | I <sub>c</sub> = 50 mA, I <sub>B</sub> = 5.0 mA   |          | 0.95        | V   |
|  |   |                        |   |          |             |     |
| T Current Gain - Bandwidth Product $I_0 = 10 \text{ mA}, V_{OE} = 20 \text{ V},$ 300 MHz $C_{obo}$ Output Capacitance $V_{CB} = 5.0 \text{ V}, I_B = 0,$ 4.0 pF $C_{0bo}$ Input Capacitance $V_{CB} = 5.0 \text{ V}, I_C = 0,$ 8.0 pF $I_{D0}$ Input Capacitance $V_{CB} = 0.5 \text{ V}, I_C = 0,$ 8.0 pF   NF Noise Figure $I_0 = 100 \mu A, V_{CE} = 5.0 \text{ V},$ 5.0 dB   SWITCHING CHARACTERISTICS $I_0 = 100 \mu A, I_{CE} = 0.5 \text{ V},$ 5.0 dB   ME Delay Time $V_{CC} = 3.0 \text{ V}, V_{BE} = 0.5 \text{ V},$ 35 ns   SwittCHING CHARACTERISTICS $I_0 = 10 mA, I_{B1} = 1.0 \text{ mA}$ 35 ns   ME Storage Time $V_{CC} = 3.0 \text{ V}, V_{BE} = 0.5 \text{ V},$ 35 ns   ME Storage Time $V_{CC} = 3.0 \text{ V}, I_C = 10 \text{ mA},$ 35 ns   ME Fail Time $I_{B1} = I_{B2} = 1.0 \text{ mA}$ 50 ns  | SMALL SIGNAL CHAR   | ACTERISTICS            |   |          | <del></del> |     |
|  | T Current Gain  | - Bandwidth Product    | Ic = 10 mA, Vcs = 20 V,<br>f = 100 MHz  | 300      |             | MHZ |
| $\begin{tabular}{ c c c c c c c c c c c c c c c c c c c$   | Cobo Output Capac   | itance                 | $V_{CB} = 5.0 \text{ V}, I_E = 0,$<br>f = 1.0 MHz   |          | 4.0         | pF  |
| Image: International system <  | Dibo Input Capacit  | ance                   | V <sub>EB</sub> = 0.5 V, I <sub>C</sub> = 0,  |          | 8.0         | pF  |
| WITCHING CHARACTERISTICS V <sub>oc</sub> = 3.0 V, V <sub>BE</sub> = 0.5 V, 35 ns   WITCHING CHARACTERISTICS V <sub>oc</sub> = 3.0 V, V <sub>BE</sub> = 0.5 V, 35 ns   Rise Time I <sub>c</sub> = 10 mA, I <sub>B1</sub> = 1.0 mA 35 ns   Storage Time V <sub>oc</sub> = 3.0 V, I <sub>c</sub> = 10mA 200 ns   Fall Time I <sub>B1</sub> = I <sub>B2</sub> = 1.0 mA 50 ns   | IE Noise Figure   |                        | f = 1.0 MHz   |          | 5.0         | dB  |
| WITCHING CHARACTERISTICS   Delay Time V <sub>CC</sub> = 3.0 V, V <sub>BE</sub> = 0.5 V, 35 ns   Rise Time I <sub>C</sub> = 10 mA, I <sub>B1</sub> = 1.0 mA 35 ns   Storage Time V <sub>CC</sub> = 3.0 V, I <sub>C</sub> = 10mA 200 ns   Fall Time I <sub>B1</sub> = I <sub>B2</sub> = 1.0 mA 50 ns   | INDISE Figure   |                        | $R_s = 1.0 \mu A$ , $v_{ce} = 5.0 v$ ,<br>$R_s = 1.0 k\Omega$ , f=10 Hz to 15.7 kHz                 |          | 5.0         | ub  |
| Delay Time V <sub>cc</sub> = 3.0 V, V <sub>BE</sub> = 0.5 V, 35 ns   Rise Time I <sub>c</sub> = 10 mA, I <sub>B1</sub> = 1.0 mA 35 ns   Storage Time V <sub>cc</sub> = 3.0 V, I <sub>c</sub> = 10mA 200 ns   Fall Time I <sub>B1</sub> = I <sub>B2</sub> = 1.0 mA 50 ns  | •   |                        | •   |          | •           |     |
| Delay Time V <sub>CC</sub> = 3.0 V, V <sub>BE</sub> = 0.5 V, 35 ns   Rise Time I <sub>c</sub> = 10 mA, I <sub>B1</sub> = 1.0 mA 35 ns   Storage Time V <sub>CC</sub> = 3.0 V, I <sub>c</sub> = 10mA 200 ns   Fall Time I <sub>B1</sub> = I <sub>B2</sub> = 1.0 mA 50 ns  | WITCHING CHARACT  | ERISTICS               |   |          |             |     |
| Rise Time Ic = 10 mA, Ig1 = 1.0 mA 35 ns   Storage Time Voc = 3.0 V, Ic = 10mA 200 ns   Fall Time Ig1 = Ig2 = 1.0 mA 50 ns   | Delay Time  |                        | V <sub>CC</sub> = 3.0 V, V <sub>BE</sub> = 0.5 V,   |          | 35          | ns  |
| Storage Time V <sub>CC</sub> = 3.0 V, I <sub>C</sub> = 10mA 200 ns   Fall Time I <sub>B1</sub> = I <sub>B2</sub> = 1.0 mA 50 ns  | Rise Time   |                        | I <sub>c</sub> = 10 mA, I <sub>81</sub> = 1.0 mA  |          | 35          | ns  |
| Fall Time I <sub>B1</sub> = I <sub>B2</sub> = 1.0 mA 50 ns   | Storage Time  |                        | V <sub>cc</sub> = 3.0 V, I <sub>c</sub> = 10mA  |          | 200         | ns  |
| *  | Fall Time   |                        | I <sub>B1</sub> = I <sub>B2</sub> = 1.0 mA  |          | 50          | ns  |
| Duise test, buise without statutes think characteristics   | * Duise Test: Duise With < 300 us   | Duty Cycle < 2.0%      |   |          |             |     |
|  |   |                        |   |          |             |     |

2N3904 / MMBT3904 / PZT3904

