

(1)

EECE 2412 - HW 6 Solutions

Prob. 1

For the circuit in Fig. 4.10 on page 224, the equation for the load line at the input is:

$$V_{BB} + V_{in}(t) = R_B \cdot i_B(t) + V_{BE}(t)$$

Substituting the given values:

$$0.8V + 0.2 \cdot \sin(2000\pi t) = 40 \times 10^3 \cdot i_B(t) + V_{BE}(t)$$

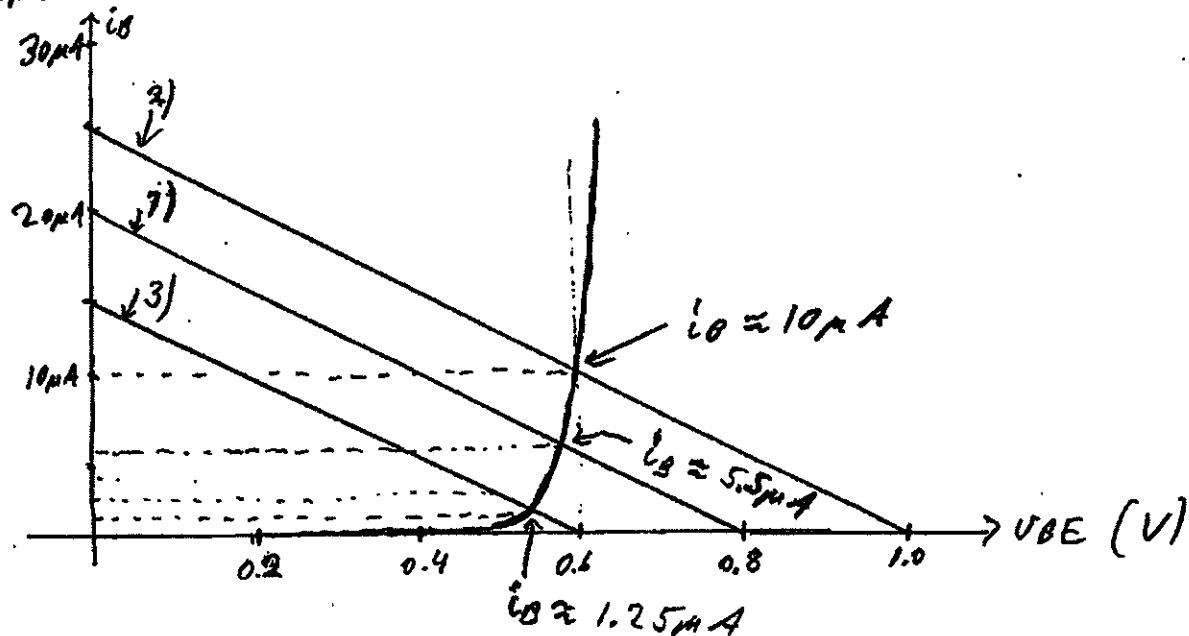
Observing the relevant cases:

$$1) V_{in}(t) = 0 : i_B = 0 \rightarrow V_{BE} = 0.8V \\ V_{BE} = 0 \text{ when } i_B = \frac{0.8V}{40000} = 20\mu A$$

$$2) V_{in}(t) = 0.2V : i_B = 0 \rightarrow V_{BE} = 1V \\ V_{BE} = 0 \text{ when } i_B = \frac{1V}{40000} = 25\mu A$$

$$3) V_{in}(t) = -0.2V : i_B = 0 \rightarrow V_{BE} = 0.6V \\ V_{BE} = 0 \text{ when } i_B = \frac{0.6V}{40000} = 15\mu A$$

Plotting the three load line cases on top of the input characteristic curve:

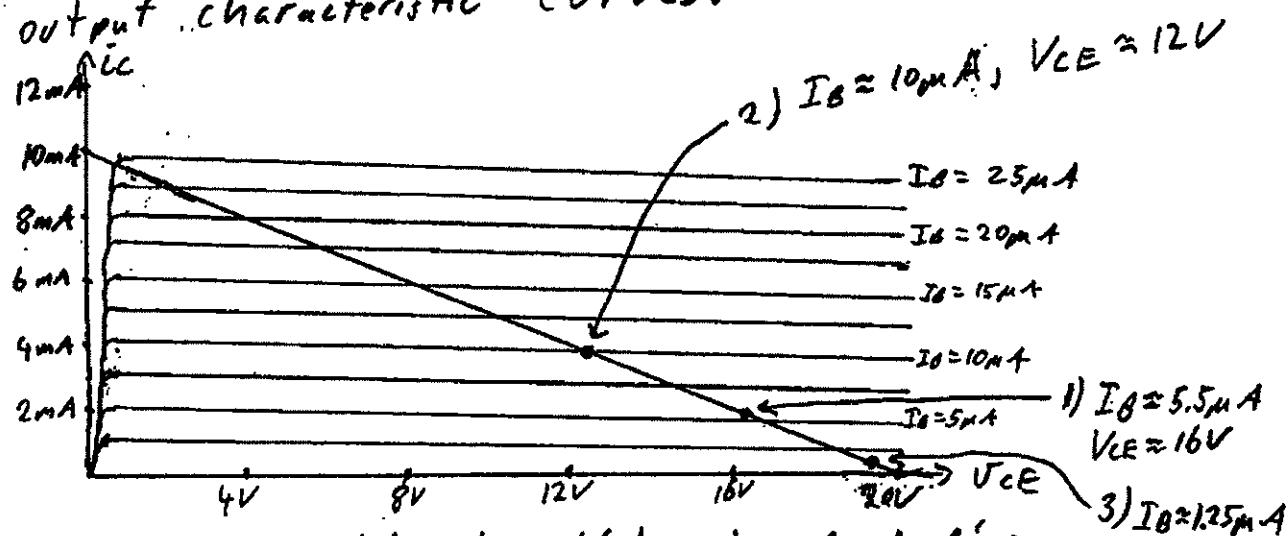


(2) ... Problem 1 continued:

Output load line equation: $V_{CC} = R_C \cdot I_C + V_{CE}$

Substituting the given values: $20 = 2000 \cdot I_C + V_{CE}$

Plotting this equation on top of the output characteristic curves:



Identify the points at which the load line intersects with the curves for the I_B values that were found on the previous page:

The min. and max. instantaneous operating points are:

$$V_{CE} \approx 12 \text{ V}, I_B \approx 10 \mu\text{A}, I_C \approx 4 \text{ mA}$$

$$V_{CE} \approx 19 \text{ V}, I_B \approx 1.25 \mu\text{A}, I_C \approx 0.5 \text{ mA}$$

The output signal is not clipped.

(Clipping would occur if $V_{CE} < 0.2 \text{ V}$ \leftarrow BJT would enter the saturation region or $V_{CE} > 20 \text{ V}$ for this case)

\hookrightarrow not possible due to the supply voltage limit

(3)

Prob. 2

$$I_c = 4 \text{ mA}$$

$$\beta = 80$$

a)

$$I_E = \frac{1+\beta}{\beta} \cdot I_c = \frac{81}{80} \cdot 4 \text{ mA} = 4.05 \text{ mA}$$

$$V_E = I_E \cdot R_E = (4.05 \times 10^{-3}) \cdot 200 = 0.81 \text{ V}$$

$$V_B = V_E + V_{BE} = 0.81 + 0.5 = 1.31 \text{ V}$$

$$I_{R2} = \frac{V_B}{R_2} = \frac{1.31}{5000} = 262 \mu\text{A}$$

$$I_B = \frac{I_c}{\beta} = \frac{4 \text{ mA}}{80} = 50 \mu\text{A}$$

$$I_{R1} = I_{R2} + I_B = 312 \mu\text{A}$$

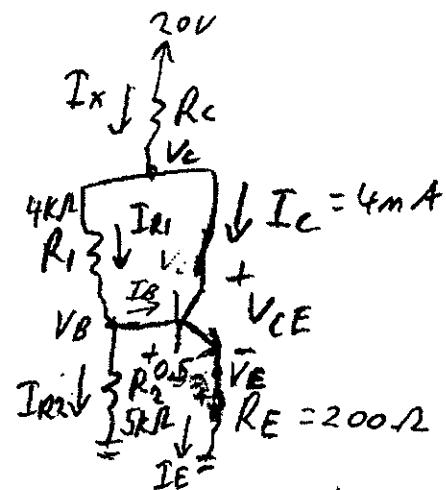
$$V_C = V_B + I_{R1} \cdot R_1 = 1.31 + (312 \times 10^{-3}) \cdot 4000 = 2.558 \text{ V}$$

$$I_X = I_c + I_{R1} = 4 \text{ mA} + 312 \mu\text{A} = 4.312 \text{ mA}$$

$$R_C = \frac{20 \text{ V} - V_C}{I_X} = \frac{20 - 2.558}{4.312 \times 10^{-3}} = \boxed{4.045 \text{ k}\Omega = R_C}$$

$$b) V_{CE} = V_C - V_E = 2.558 \text{ V} - 0.81 \text{ V}$$

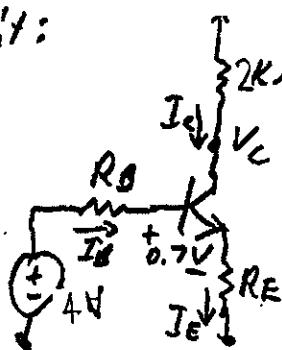
$$\boxed{V_{CE} = 1.748 \text{ V}}$$



(4)

Problem 3

a) DC circuit:



Assuming active mode (by design):

$$\textcircled{1}: I_E = \frac{I_C}{\alpha} = \left(\frac{\beta+1}{\beta}\right) \cdot I_C$$

$$\textcircled{2}: I_B = \frac{I_C}{\beta}$$

FYI:

for DC analysis:
 Capacitor \rightarrow open-circuit
 Inductor \rightarrow short-circuit

AC voltage source \rightarrow short-circuit
 AC current source \rightarrow open-circuit

$$\text{KVL: } 0 = -4V + I_B \cdot R_B + 0.7V + I_E \cdot R_E$$

$$\text{Sub. } \textcircled{1} \text{ and } \textcircled{2}: 3.3V = \left(\frac{I_C}{\beta}\right) \cdot R_B + \left(\frac{\beta+1}{\beta}\right) \cdot I_C \cdot R_E$$

case 1: Sub. $\beta = 100$, $I_C = 4mA$ into the above eqn.:

$$3.3V = \left(\frac{4mA}{100}\right) R_B + \left(\frac{100+1}{100}\right) \cdot (4mA) \cdot R_E$$

$$\textcircled{3}: 3.3V = (40 \times 10^{-6}) \cdot R_B + (4.04 \times 10^{-3}) \cdot R_E$$

case 2: Sub. $\beta = 300$, $I_C = 5mA$ into the eqn.:

$$\textcircled{4}: 3.3V = (16.667 \times 10^{-6}) \cdot R_B + (5.01667 \times 10^{-3}) \cdot R_E$$

Solving $\textcircled{3}$ and $\textcircled{4}$ simultaneously:

$$R_B = 24.1728 k\Omega$$

$$R_E = 577.5 \Omega$$

b) eqn. 4.42 in the book:

$$g_m = \frac{I_{CA}}{V_T} \implies g_m(\text{max}) = \frac{5mA}{26mV} = 0.192 \frac{A}{V} = 192 mS$$

$\hookrightarrow \approx 26mV$ at room temp. ($V_T = \frac{k \cdot T}{q}$)

$$g_m(\text{min}) = \frac{4mA}{26mV} = 153.8 mS$$

$$153.8 mS < g_m < 192 mS$$

eqn. 4.38 in the book: $r_{\pi} = \frac{\beta \cdot V_T}{I_{CA}}$

$$r_{\pi 1} = \frac{300 \times 26mV}{5mA} = 1.56 k\Omega$$

$$r_{\pi 2} = \frac{100 \times 26mV}{4mA} = 650 \Omega$$

$$650 \Omega < r_{\pi} < 1560 \Omega$$

Lecture slides: $r_{CE} = \frac{V_A}{I_{CO}}$

$$r_{CE(\text{max})} = \frac{100V}{4mA} = 25 k\Omega$$

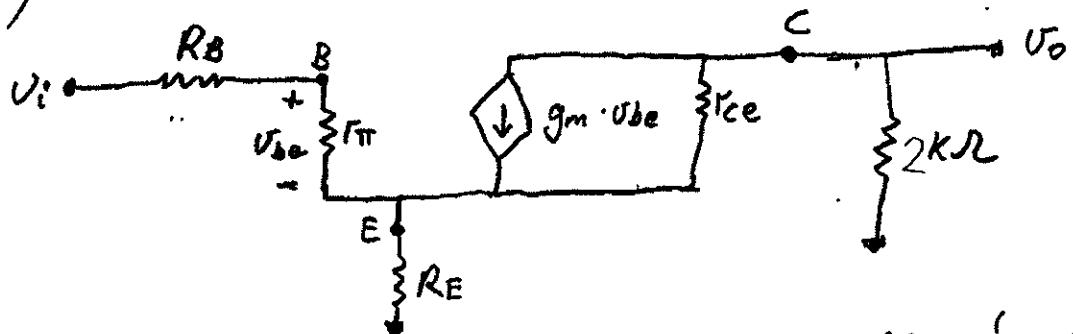
$$r_{CE(\text{min})} = \frac{100V}{5mA} = 20 k\Omega$$

$$20 k\Omega < r_{CE} < 25 k\Omega$$

4 5

... Problem 3 continued :

c) Small-signal equivalent circuit:



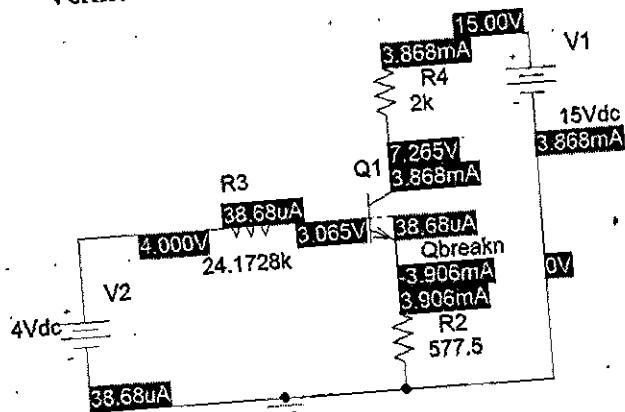
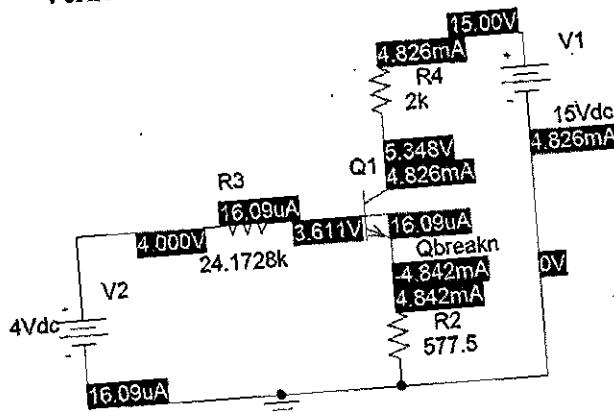
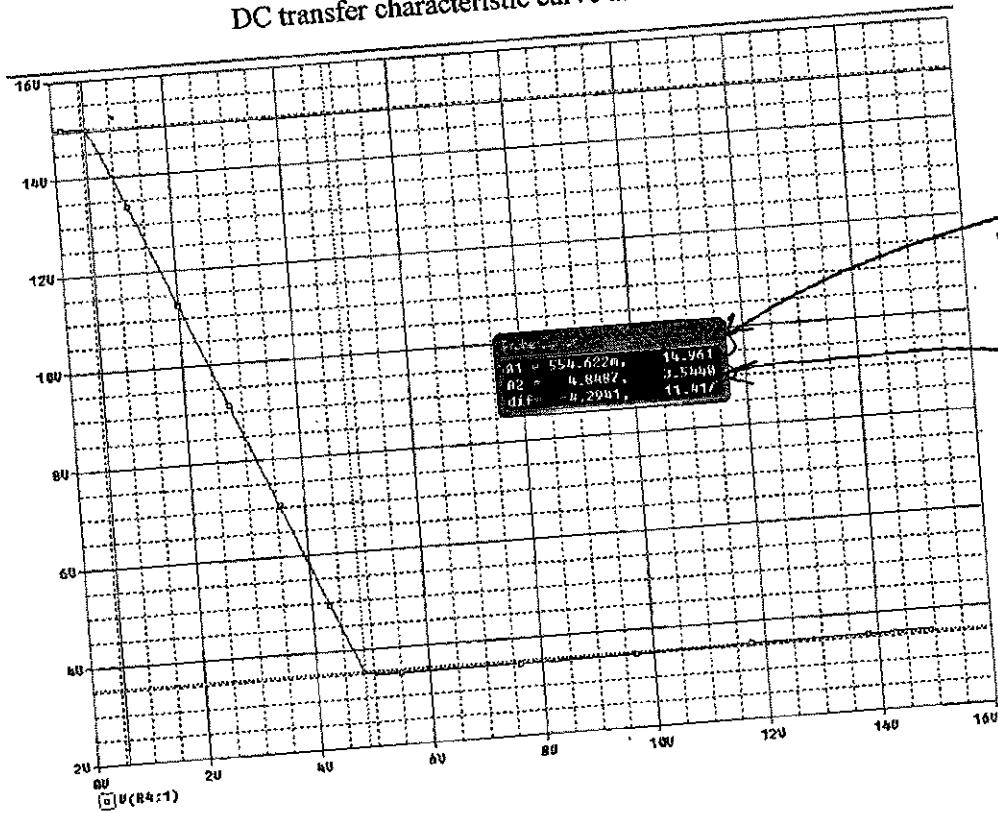
(FYI: In the AC analysis (small-signal equivalent circuit), replace:

DC voltage sources \rightarrow short-circuits
DC current sources \rightarrow open-circuits

d) PSpice: See next page

(6)

d)

Verification with $\beta=100$ (I_c is close to 4mA)Verification with $\beta=300$ (I_c is close to 5mA)DC transfer characteristic curve at the collector (V_C vs. V_{BB}) with QN222 BJT model

$V_o(\max)$
 $V_o(\max) - V_o(\min) = 11.417V$
 $= V_o(p-p)$

\Rightarrow When:
 $V_{BB} = \left(\frac{0.55 + 4.2991}{2} \right)$
 $= 2.70205V$

P.7

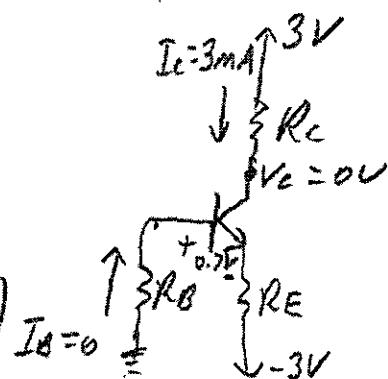
Problem 4

a) With $\beta = \infty \rightarrow I_B = 0$

$$\hookrightarrow I_E = I_C = 3mA$$

$$KVL: 0 = -3V + I_C \cdot R_C + V_C$$

$$V_C = 0V \hookrightarrow R_C = \frac{3V}{I_C} = \frac{3V}{3mA} = 1k\Omega = R_C$$



$$KVL: 0 = I_B \cdot R_B + 0.7V + I_E \cdot R_E - 3V$$

$$\hookrightarrow R_E = \frac{(3V - 0.7V)}{3 \times 10^{-3}A} = 766.67 \Omega = R_E$$

b) $\beta = 90$, constraint: $\frac{V_{RE}}{10} = V_{RB}$

$$\frac{I_E \cdot R_E}{10} = I_B \cdot R_B \rightarrow \frac{I_E R_E}{10} = \frac{I_E \cdot R_B}{\beta + 1}$$

$$\rightarrow ①: R_B = \frac{\beta + 1}{10} \cdot R_E \quad \text{using } I_E = (1 + \beta) \cdot I_B$$

$$KVL: 0 = V_{RB} + 0.7V + V_{RE} - 3V$$

$$2.3V = \frac{V_{RE}}{10} + V_{RE} \rightarrow V_{RE} = 2.0909V$$

$$I_E = \frac{I_C}{\alpha} = \frac{\beta + 1}{\beta} \cdot I_C = \frac{90 + 1}{90} \cdot 3mA = 3.0333mA$$

$$R_E = \frac{V_{RE}}{I_E} = \frac{2.0909V}{3.0333mA} = 689.3 \Omega = R_E$$

Substituting the value of R_E into equation ①:

$$R_B = \frac{90 + 1}{10} \cdot 689.3 \Omega = 6272 \Omega = R_B$$

$$R_C = \frac{3V - V_C}{I_C} = \frac{3V - 0V}{3mA} = 1k\Omega = R_C$$

(p.8) ... Problem 4 cont.:

c) Standard 5% - resistor values for the results in part b):

$$R_C = 1k\Omega$$

$$R_E = 680 \Omega$$

$$R_B = 6.2 k\Omega$$

d) With $\beta = \infty$: $I_B = 0$, $I_C = I_E$

$$\text{KVL: } 0 = I_B \cdot R_B + V_B \rightarrow V_B = 0$$

$$\text{KVL: } 0 = I_B \cdot R_B + 0.7V + V_E \rightarrow V_E = -0.7V$$

$$\text{KVL: } 0 = I_B \cdot R_B + 0.7V + I_E \cdot R_E - 3V$$

$$\hookrightarrow I_E = \frac{3V - 0.7V}{R_E} = \frac{2.3V}{680 \Omega} = 3.382 \text{ mA}$$

$$I_C = I_E = 3.382 \text{ mA}$$

With $\beta = 90$:

$$\text{KVL: } 0 = I_B \cdot R_B + 0.7V + I_E \cdot R_E - 3V$$

$$0 = \frac{I_E}{\beta+1} \cdot R_B + I_E \cdot R_E - 2.3V$$

$$2.3V = \frac{I_E}{91} \cdot 6200 + I_E \cdot 680 \rightarrow I_E = 3.074 \text{ mA}$$

$$I_C = \alpha \cdot I_E = \frac{\beta}{\beta+1} \cdot I_E = \frac{90}{91} \cdot I_E = 3.041 \text{ mA} = I_C$$

$$\text{KVL: } 0 = I_B \cdot R_B + V_B \rightarrow V_B = -I_B \cdot R_B = -\frac{I_C}{\beta} \cdot R_B = -\frac{3.041 \text{ mA}}{90} \cdot 6200 \Omega$$

$$V_B = -0.209V$$

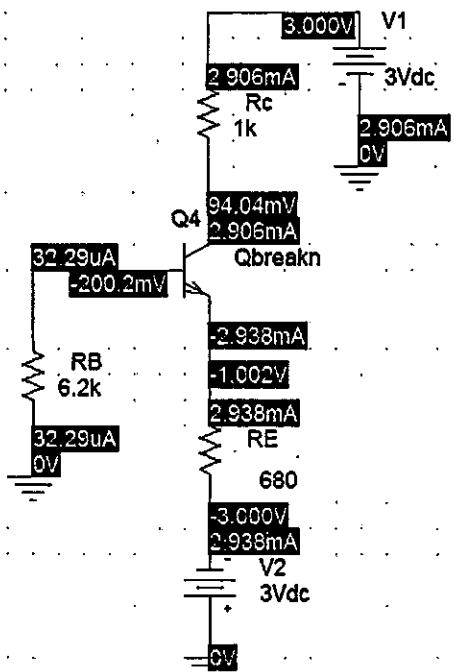
$$V_E = V_B - 0.7V = -0.909V = V_E$$

$$V_C = 3V - I_C \cdot R_C = 3V - 3.041 \text{ mA} \cdot 1k\Omega = -0.0405V = V_C$$

(P.9)

Problem 4, Part e)

with $\beta=90$:



with $\beta=400$:

