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# **ECEU692 Intro to SSI Course Notes**

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## **Part 4: Some selected topics in signal and image processing for SSI (1)**

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## Outline of those Selected Topics

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- Impulse responses and convolution (review)
- Fourier Transforms and frequency content (review)
- Convolution / Multiplication properties (review)
- Sampling and quantization and their consequences (review)
- Anti-aliasing filtering (review)
- 2-dimensional signals and all of the above
- The Projection-Slice theorem and tomography: basic idea
- What does Fourier phase tell us ?
- Power spectra and autocorrelations
- Resolution and Extent
  - apodization and its dual in time
  - the time-bandwidth product
- Relationships among varieties of Fourier representations
- Hyperspectral Images as signals

## Some conceptual stuff to put aside (for now . . . )

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- Discrete or continuous signals ?
- 1-D or 2-D or . . . ?

	Discrete	Continuous
1-D	$x(n)$	$x(t)$
2-D	$f(m, n)$	$f(x, y)$

- Discrete and continuous are conceptually very similar in how we treat them
- 1-D and 2-D are also conceptually very similar in how we treat them
- For our purposes, similarities outweigh differences
- We'll start with:
  - Discrete signals but jump back and forth as appropriate
  - 1-D, then move to 2-D

# Impulse Response and Convolution

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Let's review the 1-D discrete case

- Systems are generally time-dispersive (they can't respond instantaneously)
- We model this as an “impulse response”, the response to a very short input
- In discrete time this is the response (denoted  $h(n)$ ) to an input  $\delta(n)$  that is non-zero only at  $n = 0$
- By superposition we can find the response to any other input as a weighted sum of impulse response samples

$$y(n) = \sum_k h(k)x(n - k)$$

where the weights come from the inputs. This is the  
**Convolution Sum**

- We move in opposite directions in input and impulse response inside the sum because **later** parts of response are weighted by **earlier** inputs

Note that no limits on a sum or integral means (here) that it runs from  $-\infty$  to  $\infty$

## Frequency Content of Signals

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Some other examples besides light through a prism . . .

- Why do two instruments playing the same note (say, a violin and a clarinet) sound different ?
- Why can you immediately differentiate between the identity of speakers even if their voices have the same pitch ?
- One major reason: even though the “fundamental frequency” is the same in both cases (the “note” or the “pitch”), there are other (higher harmonic) frequencies present that differ between instruments or people
- Next question: how can you select WBCN when radio waves from WGBH are also hitting your radio receiver at the same time?
- Each station “moves” the frequency content of its music to center around its assigned “broadcast frequency”, and so that it does not overlap the frequencies used by other stations
- You select one frequency band to pass and filter out others.
- Bottom line here: The distribution of energy (and phase) with frequency is an extremely useful way to describe many signals

## Frequency Response of a System

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- The Fourier Transform of an impulse response is called the Frequency Response
- Why? It turns out that it completely describes how a system responds in the steady-state to sinusoidal inputs at each frequency:
  - The magnitude of the frequency response at a given frequency describes the gain or attenuation
  - The angle of the frequency response at a given frequency describes the delay or advance
- For 1-D discrete signals, if  $H(\omega)$  is the Discrete Time Fourier Transform (DTFT) of the impulse response  $h(n)$ , then
  - If the input is  $A \cos(\omega_0 n + \theta_0)$
  - Then the output is  $(|H(\omega_0)|A) \cos(\omega_0 n + (\theta_0 + \angle(H(\omega_0))))$
  - Note that the **frequency** does not change !!!

## Convolution and Multiplication

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- Applying the Discrete Time Fourier Transform to the Convolution Sum we get that if

$$y(n) = h(n) * x(n)$$

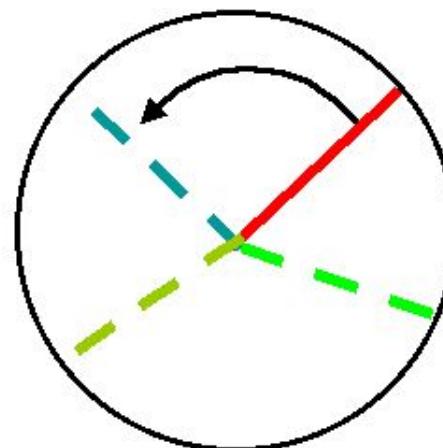
then

$$Y(\omega) = H(\omega)X(\omega).$$

- So the frequency response determines which frequencies in the input are amplified or attenuated, and how much each frequency component is delayed.
- Provides an analysis tool: multiplication is simpler than convolution
- Provides a synthesis tool: describe how you *want* a system to behave in the frequency domain and then design a filter to do it

## The Sampling Theorem: a Review

- Shannon sez: To get all the information in a signal and avoid causing distortion (aliasing), you need to sample at least twice as fast as the highest frequency present in the signal
- A simple interpretation: first consider a bar of length one performing uniform circular motion around a circle. How fast do we need to sample it to be able to know for sure where it's been all the time?



## So What's the Answer?

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- The first response is generally: once per cycle
- However this cannot distinguish opposite directions: if we sample once per second and at each sample the bar has moved  $\pi/4$  counterclockwise
  - it could be moving counterclockwise at  $1/8$  cycles per second, or
  - it could be moving **clockwise** at  $7/8$  cycles per second.
- Thus we need to sample at least **twice** per cycle.
- If we don't we get ambiguity, called **ALIASING**: for instance, if the bar is going clockwise at  $7/8$ , or counterclockwise at  $9/8$ , or clockwise at  $15/8$ , or counterclockwise at  $17/8$ , etc., cycles per second, it all looks like  $1/8$  cycle per second from the samples

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## Let's Generalize:

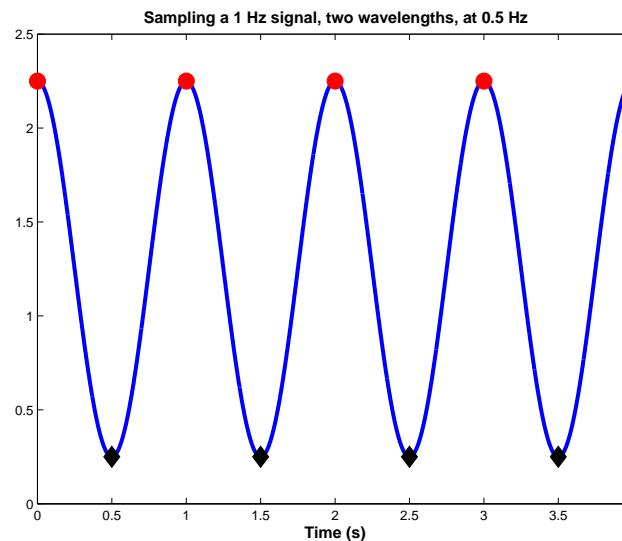
- We can write the bar's movement as  $e^{j\Omega t}$
- If we consider a superposition of "bars" of different lengths  $X(\Omega)$  moving at different (infinitely close) frequencies we get

$$\int X(\Omega) e^{j\Omega t} d\Omega$$

- This is the Fourier Transform of a signal  $x(t)$ .
- And this leads directly to Shannon's Theorem.
- Note: if the signal is *band-limited*, to a frequency band  $\Omega < B$ , this means that  $X(\Omega) = 0$  for  $\Omega \geq B$ .
- In this case, the Shannon rule says we need to sample at least as fast as  $2B/2\pi$  samples per second.

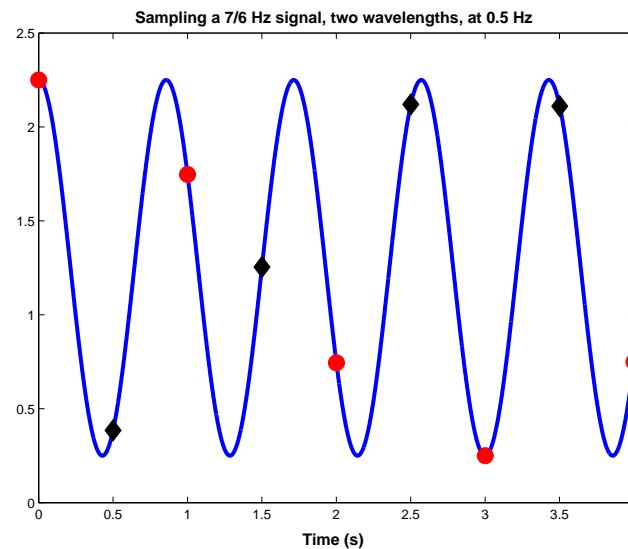
## A Simple Example

- If we are imaging blood vessels in skin, at 2 alternating wavelengths,
- We image at 30 wavelengths, and each image takes  $1/30$  second,
- The subject's pulse rate is exactly 60 beats / minute (1 Hz), then,
  - one wavelength might capture an image only when the arterial blood is at a maximum, and then
  - another would see it when it is at a minimum.
  - others might see it at various points in-between ...
  - But you certainly won't see any variation due to the pulse rate.



## Continuing the Example

- If the subject's pulse rate is 70 beats / minute ( $\approx 1.2 \text{ Hz}$ ), then
  - This aliases down to about  $0.2\text{Hz}$ , or every 5 frames, (remember the rotating bar) so
  - we will see a slow variation at each wavelength due to pulse rate



- Bottom line: if you want to see the pulse in each wavelength, you need to sample at least 2 times/pulse/wavelength for the fastest pulse you expect to see, or take fewer wavelengths, or some combination.

## Practical Sampling: Anti-Aliasing Filtering

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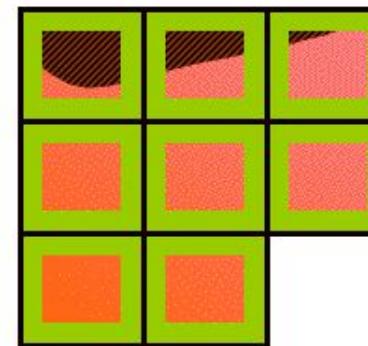
- Some practical considerations
  - No signal is truly band-limited (more on this later)
  - We generally have high-frequency noise
  - Samplers have a finite (non-instantaneous) time aperture (eg they integrate over some time period, like the CCD camera for instance)
- So:
  1. We often want to low-pass filter a signal before sampling it (called anti-aliasing filtering) to get its frequency content below the Shannon limit,
  2. Integration may help, as it's a kind of low-pass filter (details on how this works to be posted on the web site)
  3. The tradeoff is we lose *resolution* in time: everything within the integration interval gets smeared together.
  4. Sometimes we can just *integrate* in time (or space) to do this filtering

## How about “Spatial Signals” ?

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- First, let's think about a signal as a function of two spatial variables,  $f(x, y)$ , or a discrete version  $f(m, n)$ .
  - $(x, y)$  (or  $(m, n)$ ) is the location in space.
  - $f$  is the value at that location.
  - usual convention: first variable is horizontal, positive to the right, and second is vertical, positive going up.
- We can now think about filtering, for instance:
  - Low-pass in both directions, or
  - High-pass in both directions, or
  - Low-pass in one direction, High-pass in other, or
  - Even more complicated, even non-separable filters: for example, a circularly-symmetric low-pass filter.
- But what does “low-frequency”, for example, mean here ?

## Reminder: Sampling on the CCD



- The CCD camera samples this into pixels,  $f(m, n)$  by integrating over a small region on the chip.
- How do we analyze this sampling process?  
The tool we know to analyze how sampling works is “frequency content” — what does this mean for a function of space?

## Filtering in 2D

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- In the general case we have a 2D “impulse response”  $h(x, y)$  (or in the discrete case  $h(m, n)$ ), and a 2D definition of convolution (or filtering), so that the output  $g(x, y)$  relates to the input  $f(x, y)$  by

$$g(x, y) = h(x, y) * f(x, y).$$

- The discrete case looks just the same.
- $h(x, y)$  or  $h(m, n)$  are usually called the “point spread function”, 2D version of impulse response.
- Why? Well, if input is an “impulse”, that is concentrated at a single point, then the output will be  $h(x, y)$ , and will show how the system “spreads” out the impulse.
- It’s a very common and important way to characterize any imaging system — related to its spatial resolution.
- Convolution involves shifting  $h(m, n)$  over the image, multiplying, and adding (or integrating for  $h(x, y)$ .)

## Discrete Convolution in 2D

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- Let  $h(m, n)$  be a discrete psf and  $f(m, n)$  be a discrete image.
- Then the output of  $f(m, n)$  filtered by  $h(m, n)$  can be computed as

$$g(m, n) = \sum_k \sum_j f(k, j)h(m - k, n - j)$$

- Note that for each pixel position in the output,  $(m, n)$ :
  1.  $h(k, j)$  is flipped in both directions and shifted to center at  $(m, n)$
  2. The overlapping pixels of the psf and the image are multiplied and added
  3. And then you move to another position in the output image
- This is the (Double) Convolution Sum
- Usually (almost always)  $h(m, n)$  is much smaller than  $f(m, n)$  in extent, and it is often called a “mask”

## Convolution in 2D: Using a “Mask”

